

II.17 Problems

①

1) $\boxed{\text{Voltage} = \mathcal{E} = \frac{d\Phi}{dt}}$: magnetic flux $\approx \omega_0 \cdot M_0 \cdot A$

$\omega_0 = \text{Larmor}$

$M_0 = \text{equil. magnetization}$

$A = \text{Area of coil}$

* solve for a voltage which should be

$\approx 10^{-4} \rightarrow 10^{-5} \text{ volts}$

$\omega_0 = \gamma B_0$

$$M_0 = \frac{N \gamma^2 \hbar^2 B_0}{4KT} = \frac{N \gamma \hbar^2 \omega_0}{2KT}$$

$$\mathcal{E} = \frac{N \gamma \hbar^2 \omega_0 A}{2KT} = \frac{10^{20} \times 26.75 \times 10^7 \times 1.05^2 \times 10^{-68}}{2 \times 1.38 \times 10^{-23} \times 300}$$

$$\frac{300^2 \times 10^{12} \times \hbar^2 \times 0.05^2}{2 \times 1.38 \times 10^{-23} \times 300} = 8.0 \times 10^{-5} \text{ V}$$

2) The signal is: $\boxed{S_{\gamma} = \frac{\gamma^3 \hbar^2 N B_0^2}{4KT} e^{-\mathcal{E} \omega_0 t}}$

* IN CHCl_3 , The ratio of ^1H & ^{13}C is 1:1

* The Abundance of ^{13}C is 1.1%

* Let $B = \text{field for } ^{13}\text{C}$

$$1.1\% \times \gamma_{^{13}\text{C}}^3 \times B^2 = \gamma_{^1\text{H}}^3 \times 2.35^2$$

$$B = \left(\frac{\gamma_{1H}}{\gamma_{13C}} \right)^{3/2} \times \frac{1}{(1.1970)^{1/2}} \times 2.35 = 179 \text{ Tesla}$$

②

* IF molecules are ^{13}C enriched then

$$B = \left(\frac{\gamma_{1H}}{\gamma_{13C}} \right) \times 2.35 = 18.8 \text{ T}$$

3) A DBM output signal 15°

$$\textcircled{0} \quad \& \quad S_1(t) \cdot S_2(t) = \cos(\omega_1 t + \phi_1) \cdot \cos(\omega_2 t + \phi_2)$$

$$\textcircled{0} = \frac{1}{2} \cos[(\omega_1 - \omega_2)t + (\phi_1 - \phi_2)] + \frac{1}{2} \cos[(\omega_1 + \omega_2)t + (\phi_1 + \phi_2)]$$

a)

* Note output frequencies: $(\omega_1 - \omega_2)$ & $(\omega_1 + \omega_2)$

" " Phases: $(\phi_1 - \phi_2)$ & $(\phi_1 + \phi_2)$

\therefore IS sensitive TO INPUT Phases/signal

b) Let $\omega_1 = \omega_2$ & output signals are 0 & $2\omega_1$

* filter out DC & left w/ $2\omega_1$

c) IF we let $S_2 = \cos(180^\circ) = -1$, the output signals are $\cos(\omega_1 t + \phi_1 - 180^\circ)$, \therefore INPUT signal is 180° phase shifted.

d) IF we apply a low level signal at the R-PORT & the RF at L-PORT, the output has a frequency difference & lower signal level

e) when a bias current flows in the X-PORT, the DBM passes RF, when bias is OFF, no current flows.

7
0

4) IF we change the magnetic field direction "sufficiently" slowly; the magnetization follows the direction of the magnetic field.

In the rotating frame, the effective field has the components:

$$B_0 - \frac{\omega}{\gamma} \quad \} \text{ along } \hat{z}$$

B_1 is on x-y plane,

* IF the resonance frequency ω is swept slowly from well below & well above the exact resonance frequency

$$\text{eg) } \frac{\omega}{\gamma} \ll B_0 \quad \text{to} \quad \frac{\omega}{\gamma} \gg B_0,$$

We can see the component along \hat{z} change gradually from $+\hat{z}$ to $-\hat{z}$, while the component on x-y is unchanged.

7
,

5) see 170

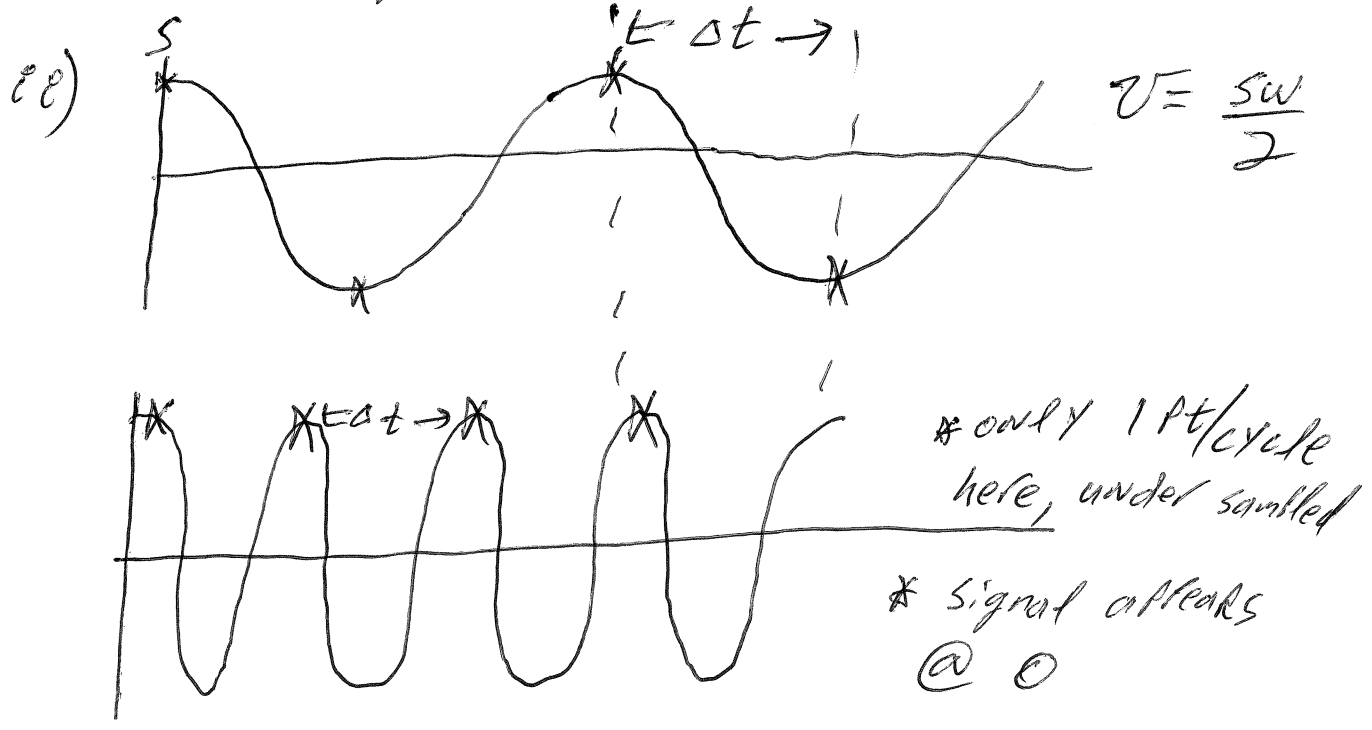
6) INVERSE FOURIER TRANSFORM IS:

$$s(t) = \int_0^{\infty} F(\omega) e^{i\omega t} d\omega$$

The Discrete Fourier Transform:

$$S_m = \sum_{k=0}^{N-1} F_k e^{i2\pi k m / N}$$

7) i) 2 pts are minimum pts needed per cycle for assigning the Resonance Frequency.



ii) $v = -s\omega$, resonance will appear at zero.

* Resonance Frequency btw $\frac{s\omega}{2}$ & $s\omega$ will appear btw $\frac{s\omega}{2}$ & zero.

* If Δt is wrong, signals appear at $\frac{1}{\Delta t} - v$ v = actual freq

8) The dwell time is:

$$\Delta t = \frac{1}{50} = \frac{1}{10\text{kHz}} = 0.1\text{ms}$$

Digital Resolution of 0.5 Hz give AT:

$$AT = \frac{1}{0.5\text{Hz}} = 2\text{s}$$

∴ Points needed are:

$$\frac{AT}{\Delta t} = \frac{2 \times 10^3}{0.1} = 2 \times 10^4 \begin{cases} \text{for FT algorithm} \\ \text{use } 32,768 \text{ pts} \end{cases}$$

$$9) \boxed{s(t) = s_0 e^{j\omega t} e^{-t/T_2} = s_0 e^{-j\pi f t} e^{-t/T_2}}$$

$$f = 1000\text{Hz}; \Delta t = 0.5\text{ms}; \frac{1}{\pi T_2} = 1\text{Hz}; T_2 = 0.318\text{s}$$

$$S_n = s_0 e^{-j2\pi f n \Delta t} e^{-n \Delta t / T_2} = e^{-j\pi n} e^{-n \times 1.57 \times 10^{-3}}$$

$$S_0 = s_0 \cdot 1$$

$$S(1) = s_0 e^{-j\pi} e^{-1 \times 0.5 \times 10^{-3} / 0.318}$$

$$= \cancel{s_0 0.9969} = 0.998 \cdot s_0$$

$$S(2) = s_0 e^{-2\pi} e^{-2.14 \times 10^{-3}} = \cancel{0.994 s_0}$$

$$0.9969 s_0$$

$$S(3) = s_0 e^{-3\pi} e^{-4.11 \times 10^{-3}} = 0.994 \cdot s_0$$

↓ and so on

10) After a $T/2$ pulse, the magnetization is on the x-y plane. Afterwards the magnetization goes as:

$$M_{x,y} = M_0 e^{-t/T_2}$$

A magnetization along z-axis is:

$$M_z = (1 - e^{-t/T_1}) M_0$$

∴ Total length M :

$$M = \sqrt{M_{x,y}^2 + M_z^2} = M_0 \sqrt{e^{-2t/T_2} + (1 - e^{-t/T_1})^2}$$

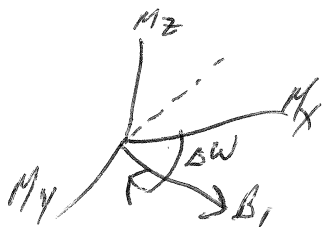
$$= M_0 \sqrt{1 - 2e^{-t/T_1} + e^{-2t/T_2} + e^{-2t/T_2}}$$

$$M = M_0 (1 - 2e^{-t/T_1} + 2e^{-2t/T_2})^{1/2} \quad \text{where } T_1 = T_2$$

$$11) \frac{dM_x}{dt} = \gamma (M_y B_0 + M_z B_1 \sin \omega t) - \frac{M_x}{T_2}$$

$$\frac{dM_y}{dt} = \gamma (M_z B_1 \cos \omega t - M_x B_0) - \frac{M_y}{T_2}$$

$$\frac{dM_z}{dt} = \gamma (-M_x B_1 \sin \omega t - M_y B_1 \cos \omega t) - \frac{(M_z - M_0)}{T_1}$$



$$\omega = -\gamma B_0$$

$$(2) F(\omega) = \frac{1}{T_2} \int_{-\infty}^{\infty} s(t) e^{-i\omega t} dt$$

$$= \frac{1}{T_2} \int_{-\infty}^{\infty} e^{i\Delta\omega t - t/T_2} e^{-i\omega t} dt$$

$$\text{Recall: } s(t) = e^{i\Delta\omega t - t/T_2}$$

$$= \frac{1}{T_2} \int_{-\infty}^{\infty} e^{(i\Delta\omega - i\omega - 1/T_2)t} dt$$

$$= \frac{1}{T_2} \frac{1}{i(\Delta\omega - \omega) - 1/T_2} = \frac{1}{T_2} \frac{i(\Delta\omega - \omega) + 1/T_2}{-(\Delta\omega - \omega)^2 - 1/T_2^2}$$

$$F(\omega) = \frac{T_2}{1 + (\omega - \Delta\omega)^2 T_2^2} + \frac{2T_2^2 (\omega - \Delta\omega)}{T_2^2 (\omega - \Delta\omega)^2 + 1}$$

\Downarrow ABSORPTIVE \Downarrow DISPERSIVE

(a) $A(\omega) = \frac{1}{1 + (\omega - \Delta\omega)^2 T_2^2}$ when $\omega = \Delta\omega$ $A(\omega)$ has a maximum

(b) $A(\omega) = 1$
 The Half-height will be $A(\omega) = 1/2$

$$\frac{1}{1 + (\omega - \Delta\omega)^2 T_2^2} = 1/2$$

$$T_2^2 (\omega - \Delta\omega)^2 = 1 \Rightarrow \omega_{\pm} = \Delta\omega \pm 1/T_2$$

$$\therefore \text{FWHM} = \omega_+ - \omega_- = \frac{2}{T_2}$$

(8)

18) The magnitude is: $\sqrt{A(\omega)^2 + D(\omega)^2} = \frac{1}{\sqrt{1 + (\omega - \omega_0)^2 T_0^2}}$

When $\omega = \omega_0$, have $\text{max} = 1$

∴ at FWHH: $\frac{1}{\sqrt{1 + (\omega - \omega_0)^2 T_0^2}} = \frac{1}{2}$

$$(\omega - \omega_0)^2 T_0^2 = 3$$

$$\omega_{\pm} - \omega_0 = \pm \frac{\sqrt{3}}{T_0}$$

$$\omega_{\pm} = \omega_0 \pm \frac{\sqrt{3}}{T_0}$$

∴ The FWHH $\Rightarrow \omega_+ - \omega_- = \frac{2\sqrt{3}}{T_0}$

* comparison of Magnitude vs. Phased FWHH reveals the Phased is narrower by a factor of $\sqrt{3}$!

14) Phase collection eqn: $\phi = \theta + \Delta\omega \cdot \tau$

ON resonance ∴ $\Delta\omega = 0$

$$\phi = 45^\circ$$

FROM sequence: $\tau = 50 \mu\text{s} + 100 \mu\text{s} = 150 \mu\text{s}$

i) $\Delta\omega = 5 \text{ kHz}$

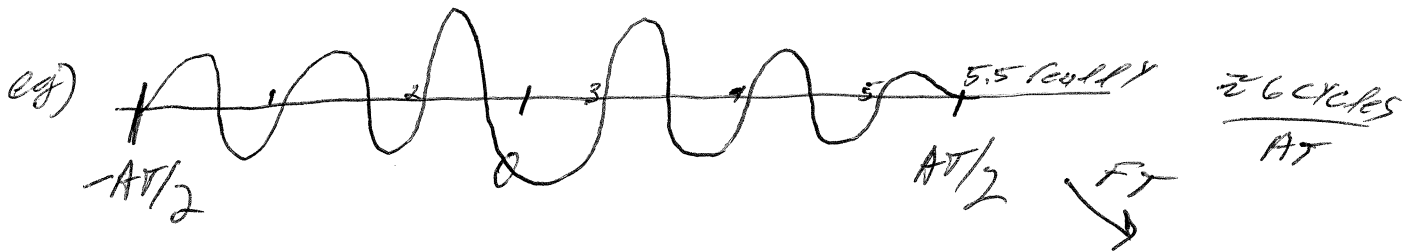
$$\phi = 45^\circ + 5 \text{ kHz} \times 150 \mu\text{s} \times \left(\frac{360^\circ}{\text{cycle}} \right) = 315^\circ$$

14) eg) $\Delta\omega = -5\text{KHz}$

$$\phi = 45^\circ - 5\text{KHz} \times 150\text{ms} \times \frac{360^\circ}{\text{cycle}} = -225^\circ$$

15) The $AT = 1024 \times 500\text{ms} = 512,000\text{ms}$

The Resonance Freq. = $\frac{6}{512,000\text{ms}} = 11\text{Hz}$



* Assume both TR & II signal is collected

* 180° Phase correction needed to get $A(\omega)$

* By using linear phase correction given by

$\tau = AT/2$, like a dead time correction we get $A(\omega)$

$$\therefore \tau = \frac{180^\circ}{11.7 \times 360^\circ} = 42.7\text{ms}$$

OR IS IT $\left[\phi = 180 - \frac{512,000\text{ms} \times 11.7 \times 360^\circ}{2} \right]$

THINK ITS THIS REALLY ↗

(16) Since signal is linearly proportional to sample concentration & $\sqrt{\quad}$ dependent on noise:

$$\frac{10}{1 \times \sqrt{1}} = \frac{100}{0.2 \times \sqrt{\text{scans}}} \Rightarrow \text{scans} = 2500$$

(17) Let's use t_- & t_+ to represent time before & after the pulse β .

Z-Magnetization after pulse is:

$$M_z(t_+) = M_z(t_-) \cos(\beta)$$

Z-magnetization after acquisition AT:

$$M_z(AT) = M_z(t_+) e^{-AT/T_2} + M_0 (1 - e^{-AT/T_1})$$

Since $T_1 = T_2$

$$M_z(AT) = M_z(t_+) e^{-AT/T_1} + M_0 (1 - e^{-AT/T_1})$$

When equilibrium is reached: $M_z(AT) = M_z(t_-)$

$$M_z(t_-) = M_0 \frac{1 - e^{-AT/T_1}}{1 - e^{-AT/T_1} \cos(\beta)}$$

however, detect $M_{x,y}$ projection

$$M_{x,y}(t_+) = M_0 \left(\frac{1 - e^{-AT/T_1}}{1 - e^{-AT/T_1} \cos(\beta)} \right) \sin(\beta)$$

cont., we have the max at an angle when

$$\frac{dM_{x,y}(\theta_+)}{d\beta} = 0$$

$$\therefore \frac{dM_{x,y}(\theta_+)}{d\beta} = M_0(1 - e^{-AT/T_1}) \left[\frac{\cos(\beta)}{1 - e^{-AT/T_1} \cos(\beta)} - \right.$$

$$\left. \frac{e^{-AT/T_1} \sin^2(\beta)}{(1 - e^{-AT/T_1} \cos(\beta))^2} \right] = \frac{M_0(1 - e^{-AT/T_1})(\cos\beta - e^{-AT/T_1})}{1 - e^{-AT/T_1} \cos(\beta)} = 0$$

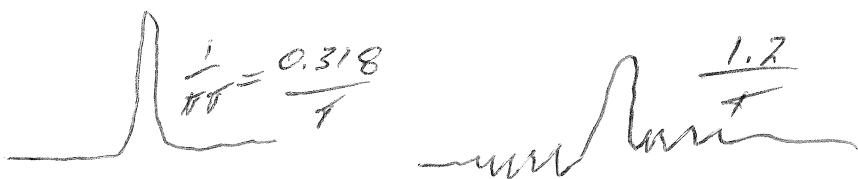
$$\therefore \boxed{\cos(\beta) = e^{-AT/T_1}} \quad \checkmark$$

(8) Let T = acquisition time for a fully collected FID.

$$\therefore fwhh = \frac{1}{\pi T_2} = \frac{0.318}{T_2}$$

* If we collect only $T/4$ we get $\frac{\sin(x)}{x} = \text{sinc } x$

$$\therefore fwhh = \frac{1}{\pi T_2} = \frac{0.318}{T/4} = \frac{1.2}{T}$$



* can use LB to smooth out sinc, but loose resolution

19) create a vortex, effects qhims/susceptibility (12)

20) R: $s(t) = \sin(\omega t) = \frac{-i(e^{i\omega t} - e^{-i\omega t})}{2}$

$$\left. \begin{array}{l} e^{i\omega t} \Rightarrow A^+ + D^+ \\ e^{-i\omega t} \Rightarrow A^- + D^- \end{array} \right\} \frac{i(e^{-i\omega t} - e^{i\omega t})}{2}$$

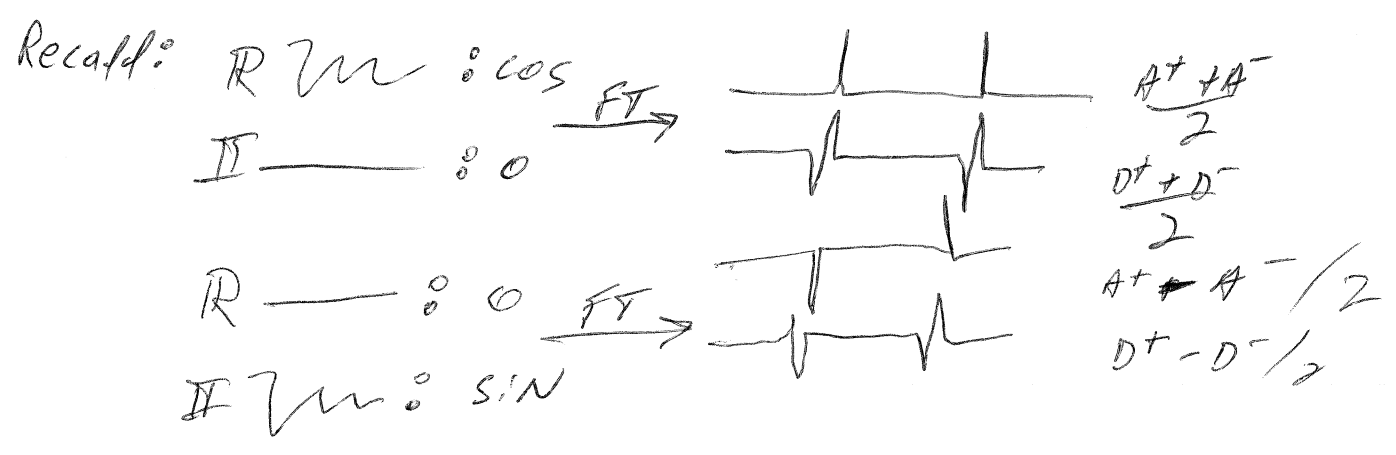
$$\Downarrow \\ \frac{i(A^- + D^- - A^+ - D^+)}{2}$$

$$F(\omega) = i \left(\frac{A^- - A^+}{2} \right) + \frac{D^+ - D^-}{2}$$

$$\begin{aligned} i) s(t) &= i \cos(\omega t) = i \frac{(e^{i\omega t} + e^{-i\omega t})}{2} \\ &= i \frac{(A^+ + D^+ + A^- + D^-)}{2} \end{aligned}$$

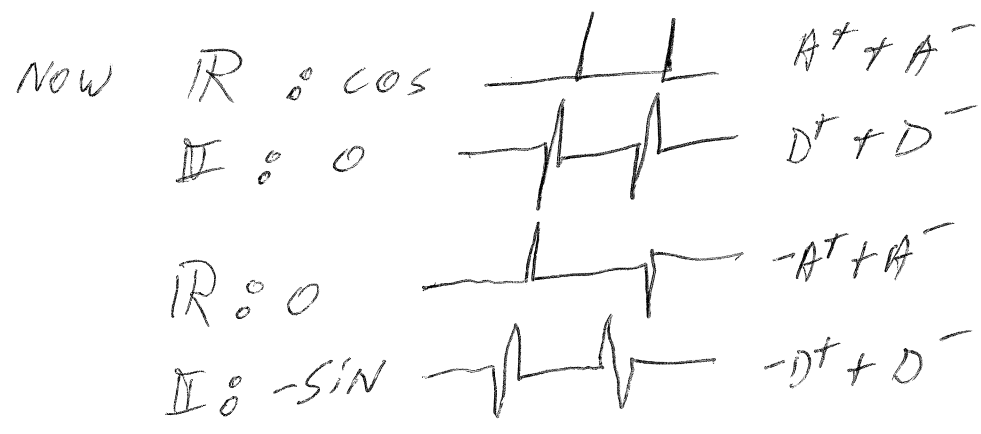
$$F(\omega) = i \left(\frac{A^+ + A^-}{2} \right) - \frac{D^+ + D^-}{2}$$

21) $F(\omega) = A(\omega) + i D(\omega) \Rightarrow \text{detects } I$



* FOR detecting I_+ or component oscillating @ $\Delta\omega$, simply acquire $-\sin$ IN I .

$\therefore \boxed{F(\omega) = A(\omega) - i D(\omega)}$ [can show w/o a.m. too]



* Do w/ Phase Cycling & Receiver (ADC buffers)

expt	ADC1 (R)	ADC2 (I)	Buffer (1)	Buffer (2)
1	cos	sin	ADC1	-ADC2
2	-sin	cos	ADC2	ADC1
3	-cos	-sin	-ADC1	ADC2
4	sin	-cos	-ADC2	-ADC1

} $4\cos$
 $-4\sin$

22) e) DYNAMIC Range of a 12-bit Digitizer:

$$\frac{2^{12}}{2} = 2^{11} = 2048$$

DYNAMIC Range of a 16-bit Digitizer

$$\frac{2^{16}}{2} = 2^{15} = 32768$$

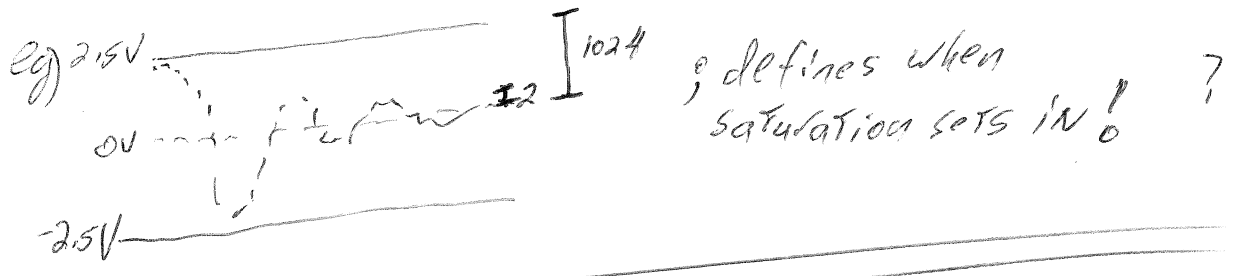
ii) The maximum signal voltage should be set close to 5V, but not to exceed,

iii) a) $S/N = \infty$ ∴ # of scans needed is:

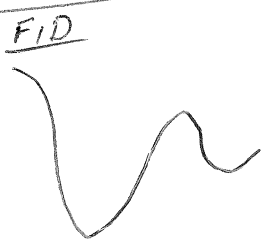
$$\frac{10^{23}}{10^{15}} = 256$$

b) $S/N \neq \infty$ ∴ # of scans needed is:

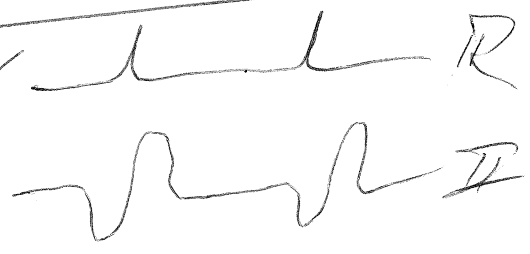
$$\left(\frac{10^{23}}{10^{15}}\right)^2 = 65,536$$



23) Larger signal IR



FT →



smaller signal IR

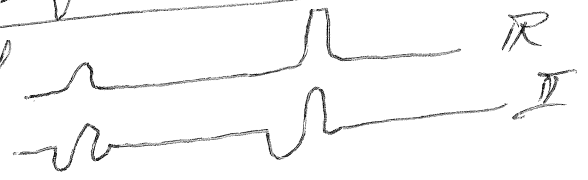


FT →



ghost intensity: $\frac{SB}{2}$

add



(15)

ϕ_{rx}	IR signal	IF signal
0	$C(\Delta\omega t)$	$S(\Delta\omega t)$
90°	$-S(\Delta\omega t)$	$C(\Delta\omega t)$
180°	$-C(\Delta\omega t)$	$-S(\Delta\omega t)$
270°	$S(\Delta\omega t)$	$-C(\Delta\omega t)$

* see Pg 63 regarding phase factor ϕ

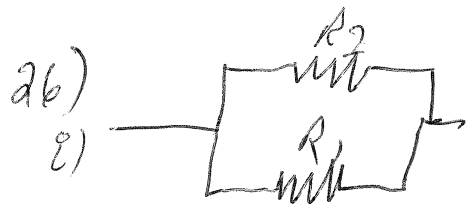
25) $\phi_{rx} = \phi_{rx}^0 + \frac{2\pi(\ell-1)}{N}$; $\ell = 1, 2, \dots, N$
 $\phi_{rx}^0 = 0 \rightarrow 360^\circ$

$C_1^\ell = \cos[2\pi(\ell-1)/N]$ $C_2^\ell = -\sin[2\pi(\ell-1)/N]$

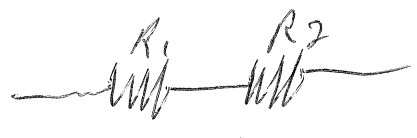
$C_3^\ell = \sin[2\pi(\ell-1)/N]$ $C_4^\ell = \cos[2\pi(\ell-1)/N]$

Let $N=4$ (cycles)

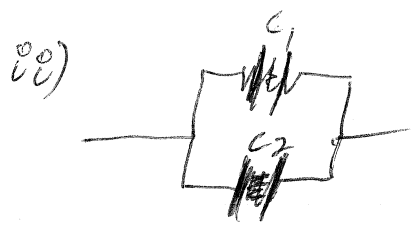
$\ell=1$	$\ell=2$	$\ell=3$	$\ell=4$	wrong
$C_1^1 = 1$	0	-1	1	1
$C_2^1 = 0$	-1	0	0	
$C_3^1 = 0$	1	0	0	
$C_4^1 = 1$	0	-1	1	



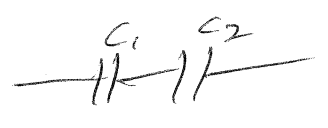
$$R = \frac{R_1 R_2}{R_1 + R_2} \quad Z = \frac{R_1 R_2}{R_1 + R_2}$$



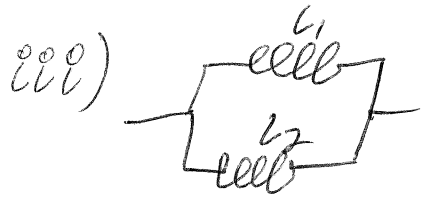
$$R = R_1 + R_2 \quad Z = R_1 + R_2$$



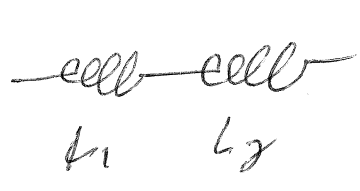
$$C = C_1 + C_2 \quad Z = \frac{-j}{\omega C} = \frac{-j}{\omega(C_1 + C_2)}$$



$$C = \frac{C_1 C_2}{C_1 + C_2} \quad Z = \frac{-j}{\omega C} = \frac{-j(C_1 + C_2)}{\omega C_1 C_2}$$



$$L = \frac{L_1 L_2}{L_1 + L_2} \quad Z = j\omega L = j\omega \left(\frac{L_1 L_2}{L_1 + L_2} \right)$$

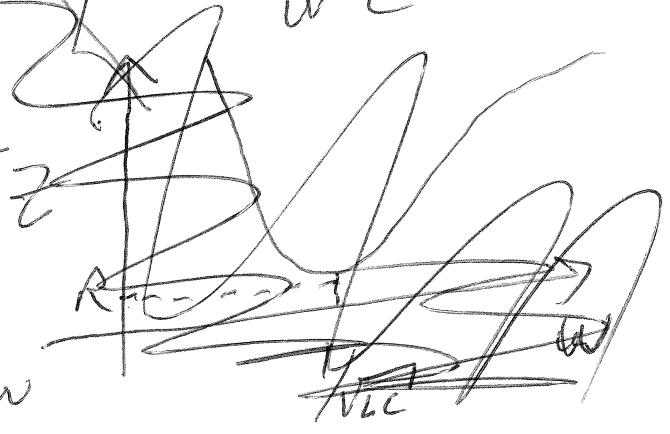
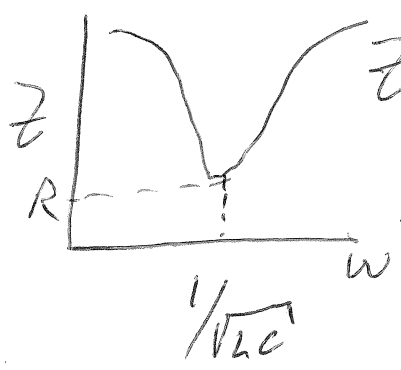


$$L = L_1 + L_2 \quad Z = j\omega(L_1 + L_2)$$

27) i) $Z_{eq} = R + j\omega L + \frac{j}{\omega C}$ $Z_{eq} = \sqrt{R^2 + (j\omega L + \frac{j}{\omega C})^2}$

27) ii) $\frac{dZ}{d\omega} = jL + \frac{j}{\omega^2 C} = 0$

$$X_T = \frac{j\omega L}{\omega^2 C L^2 + 1}$$



$$Z = \frac{R \cdot X_T}{\sqrt{R^2 + X_T^2}}$$

?

29) The output power is:

$$Z_s = -Z_L \quad \text{shouldn't it be}$$

(16)

$$P_{out} = \frac{V^2 Z_L}{(Z_s + Z_L)^2} = \frac{V^2}{\left(\frac{Z_s}{\sqrt{Z_L}} + \sqrt{Z_L}\right)^2}$$

30) i) Ideally $V = 0$

ii) $V = 0$

31) Let $Z_1 =$ impedance of tuning circuit

$$\therefore \frac{1}{Z_1} = i\omega C_2 + \frac{1}{r + i\omega L}$$

$$\text{Note: } (a+ib)(a-ib) = a^2 + b^2$$

$$= \frac{r}{r^2 + \omega^2 L^2} + i \left[\omega C_2 - \frac{\omega L}{r^2 + \omega^2 L^2} \right]$$

$$\therefore Z_1 = \frac{r/(r^2 + \omega^2 L^2) - i \left[\omega C_2 - \frac{\omega L}{r^2 + \omega^2 L^2} \right]}{r^2/(r^2 + \omega^2 L^2)^2 + \left[\omega C_2 - \frac{\omega L}{r^2 + \omega^2 L^2} \right]^2}$$

The total impedance is:

$$Z = Z_1 + \frac{1}{i\omega C_1}$$

at the resonance condition, II part is zero

(19)

$$\text{II} \approx \frac{[wC_2 - wL/(r^2 + w^2L^2)]}{r^2/(r^2 + w^2L^2)^2 + [wC_2 - wL/(r^2 + w^2L^2)]} - \frac{1}{wC_1} = 0$$

$$R \approx \frac{r/(r^2 + w^2L^2)}{r^2/(r^2 + w^2L^2)^2 + [wC_2 - wL/(r^2 + w^2L^2)]^2} = 50 \Omega$$

From R

$$C_2 = \frac{L}{(r^2 + w^2L^2)} \pm \frac{\left[\frac{r}{50 \Omega} - \frac{r^2}{r^2 + w^2L^2} \right]^{1/2}}{r^2 + w^2L^2}$$

Since: $r \ll wL$

$$C_2 = \frac{1}{w^2L} \pm \frac{\left(\frac{r}{50 \Omega} \right)^{1/2}}{w^2L}$$


$$\therefore C_2 = \frac{1 \pm \sqrt{r/50 \Omega}}{w^2L}$$

From II: $\frac{wC_2 - 1/wL}{[wC_2 - 1/wL]^2} + \frac{1}{wC_1} = 0$ again: $r \ll wL$ Trick

$$C_1 = \frac{1}{\omega} \sqrt{\frac{1}{50 \Omega}} / \omega^2 L$$

Since $C_1 > 0$, substitute C_2 & solve for C_1 .

$$C_1 = \frac{\sqrt{\frac{1}{50 \Omega}}}{\omega^2 L} \quad \& \quad C_2 = \frac{1 - \sqrt{\frac{1}{50 \Omega}}}{\omega^2 L}$$

32) When the Tx is ON both cross diodes are "on" since the voltage is $> 0.5 V$. At one end of $\lambda/4$ (right before ground) V_f is maximum & at the other end it's zero. Essentially there's no voltage drop thru the $\lambda/4$ & so all current & power is delivered to probe circuit. When Tx is OFF, $\lambda/4$ power falls below $0.5 V$ & diodes act like  open circuit. Now spin induced voltages ($10^{-5} V$) travel smoothly to Rx.

33) e) magnetization is $-M_z$

e) magnetization is in $z-y$ plane & has an α angle = $\frac{19}{20}$ to z .

(2)

$$33) \text{ i) } I_z \xrightarrow{\left(\frac{19\pi}{20}\right)} I_z c\left(\frac{19\pi}{20}\right) - I_y s\left(\frac{19\pi}{20}\right)$$

$$= -0.9876 I_z - 0.156 I_y$$

$$\text{ii) } I_z \xrightarrow{\left(\frac{19\pi}{40}\right)} I_z c\left(\frac{19\pi}{40}\right) - I_y s\left(\frac{19\pi}{40}\right) \xrightarrow{\left(\frac{19\pi}{20}\right)}$$

$$\left[I_z c\left(\frac{19\pi}{20}\right) + I_y s\left(\frac{19\pi}{20}\right) \right] c\left(\frac{19\pi}{40}\right) - I_y s\left(\frac{19\pi}{40}\right) \xrightarrow{\left(\frac{19\pi}{40}\right)}$$

$$I_z c^2\left(\frac{19\pi}{40}\right) c\left(\frac{19\pi}{20}\right) - I_y s\left(\frac{19\pi}{40}\right) c\left(\frac{19\pi}{40}\right) c\left(\frac{19\pi}{20}\right)$$

$$+ I_x s\left(\frac{19\pi}{20}\right) c\left(\frac{19\pi}{40}\right) - I_y c\left(\frac{19\pi}{40}\right) s\left(\frac{19\pi}{40}\right) - I_z s\left(\frac{19\pi}{20}\right) s\left(\frac{19\pi}{40}\right)$$

$$= -0.9998 I_z - 0.06095 I_y + 0.0520 I_x$$



Much better I_z inversion vs. i)